

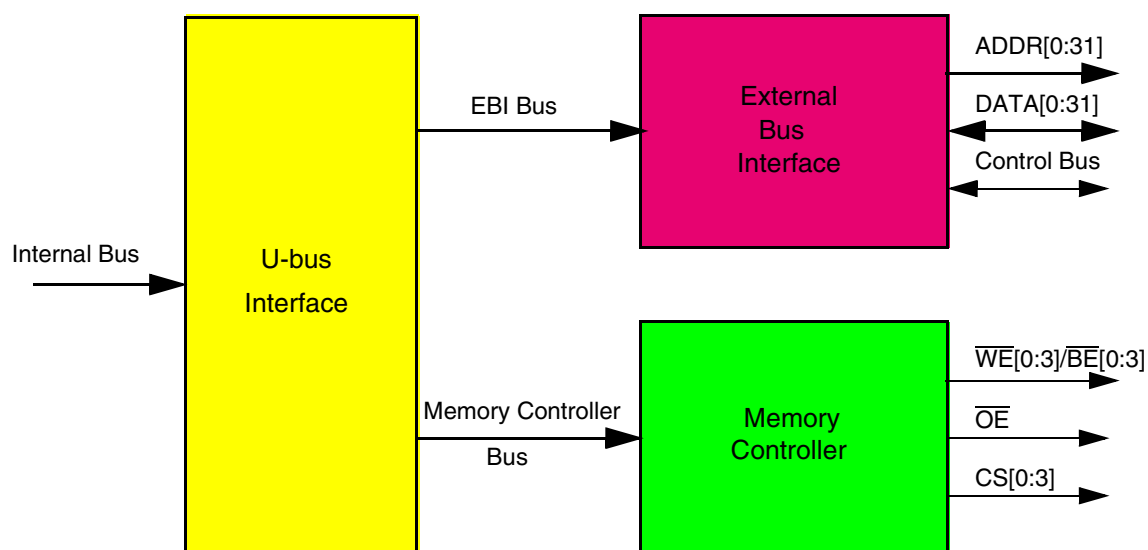


## SECTION 10

### MEMORY CONTROLLER

The memory controller generates interface signals to support a glueless interface to external memory and peripheral devices. It supports four regions, each with its own programmed attributes. The four regions are reflected on four chip-select pins. Read and write strobes are also provided.

The memory controller operates in parallel with the external bus interface to support external cycles. When an access to one of the memory regions is initiated, the memory controller takes ownership of the external signals and controls the access until its termination. Refer to [Figure 10-1](#).



**Figure 10-1 Memory Controller Function Within the USIU**

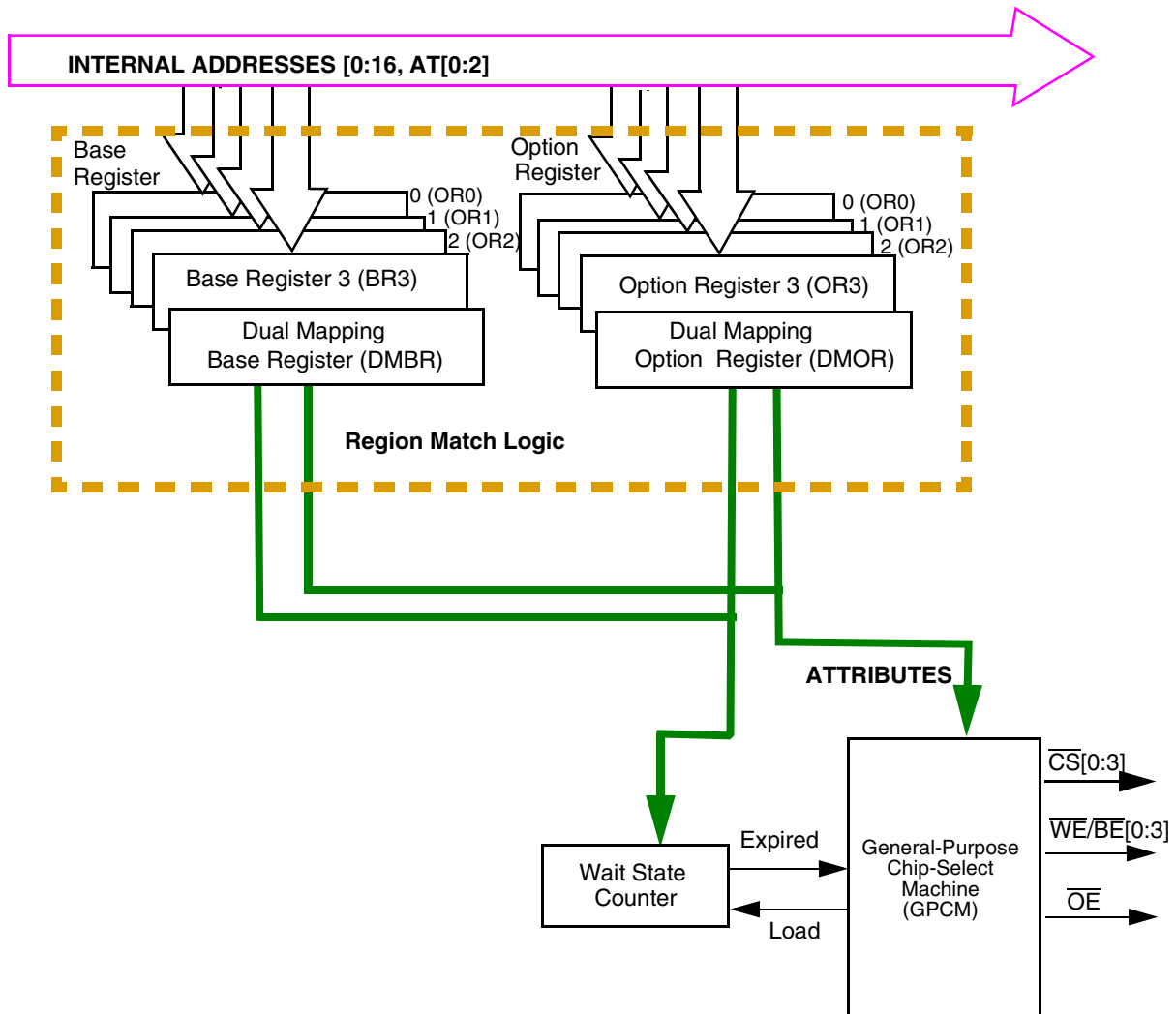
#### 10.1 Overview

The memory controller provides a glueless interface to EPROM, static RAM (SRAM), Flash EPROM (FEPROM), and other peripherals. The general-purpose chip-selects are available on lines  $\overline{CS}[0]$  through  $\overline{CS}[3]$ .  $\overline{CS}[0]$  also functions as the global (boot)

chip-select for accessing the boot flash EEPROM. The chip select allows zero to 30 wait states.



**Figure 10-2** is a block diagram of the MPC555 / MPC556 memory controller.



**Figure 10-2 Memory Controller Block Diagram**

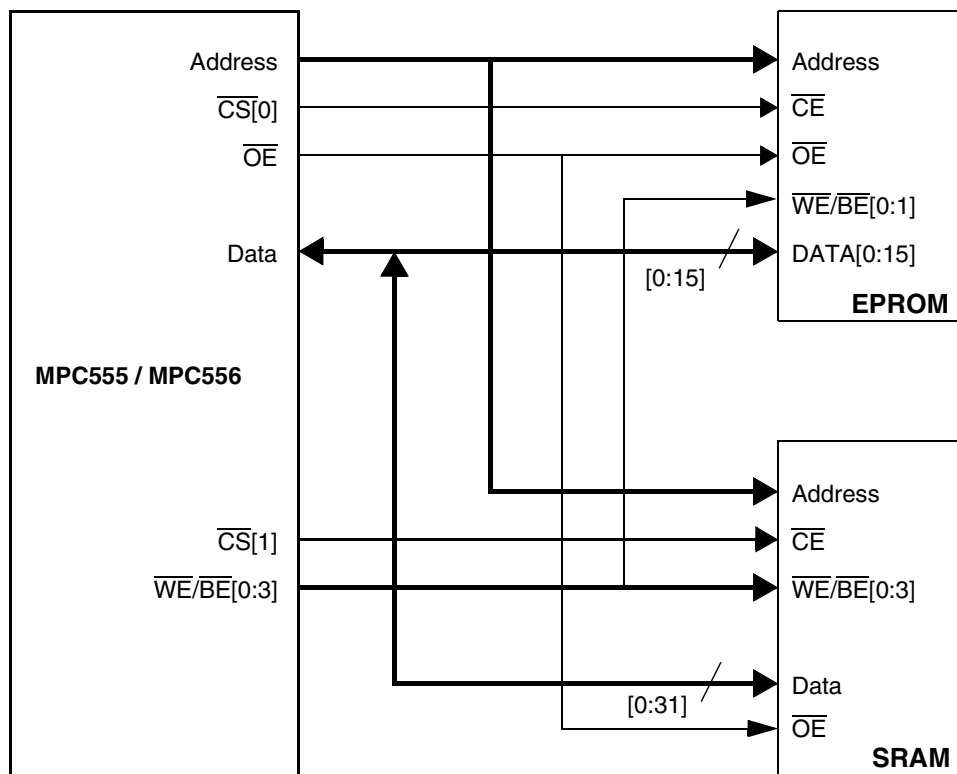
Most memory controller features are common to all four banks. (For features unique to the  $\overline{CS}[0]$  bank, refer to [10.4 Global \(Boot\) Chip-Select Operation](#).) A full 32-bit address decode for each memory bank is possible with 17 bits having address masking. The full 32-bit decode is available, even if all 32 address bits are not sent to the MPC555 / MPC556 pins.

Each memory bank includes a variable block size of 32 Kbytes, 64 Kbytes and up to 4 Gbytes. Each memory bank can be selected for read-only or read/write operation. The access to a memory bank can be restricted to certain address type codes for system protection. The address type comparison occurs with a mask option as well.

From zero to 30 wait states can be programmed with  $\overline{TA}$  generation. Four byte-write and read-enable signals ( $\overline{WE}/\overline{BE}[0:3]$ ) are available for each byte that is written to memory. An output enable ( $\overline{OE}$ ) signal is provided to eliminate external glue logic. A memory transfer start ( $\overline{MTS}$ ) strobe permits one master on a bus to access external memory through the chip selects on another.



The memory controller functionality allows MPC555 / MPC556-based systems to be built with little or no glue logic. A minimal system using no glue logic is shown in **Figure 10-3**. In this example  $\overline{CS}[0]$  is used for the 16-bit boot EPROM and  $\overline{CS}[1]$  is used for the 32-bit SRAM. The  $\overline{WE}/\overline{BE}[0:3]$  signals are used both to program the EPROM and to enable write access to various bytes in the RAM.

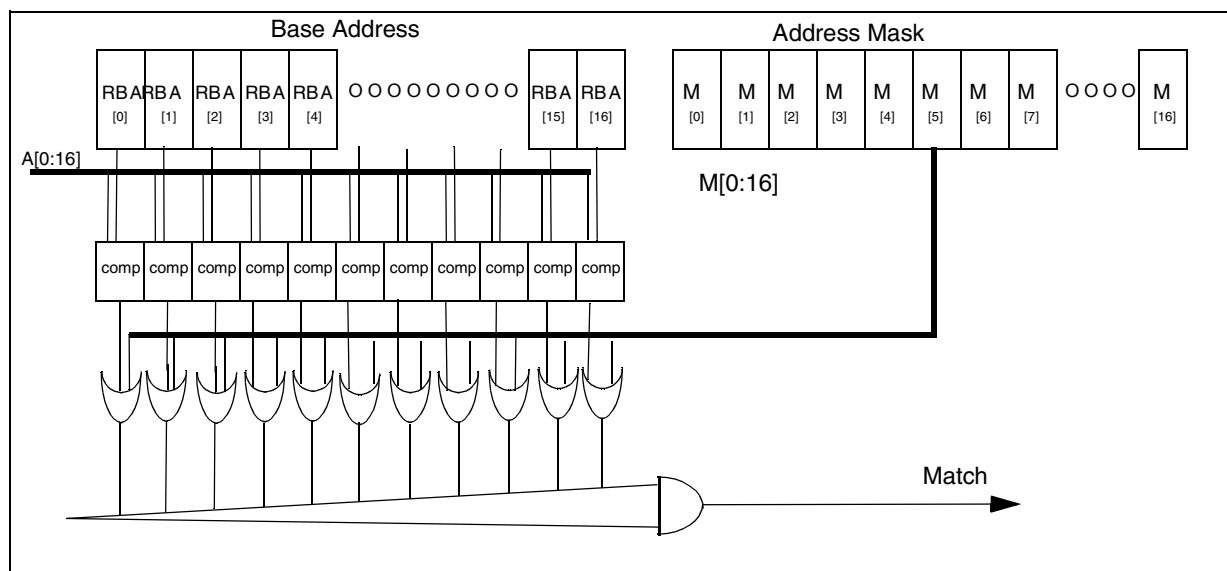


**Figure 10-3 MPC555 / MPC556 Simple System Configuration**

## 10.2 Memory Controller Architecture

The memory controller consists of a basic machine that handles the memory access cycle: the general-purpose chip-select machine (GPCM).

When a new access to external memory is requested by any of the internal masters, the address of the transfer (with 17 bits having mask) and the address type (with 3 bits having mask) are compared to each one of the valid banks defined in the memory controller. Refer to **Figure 10-4**.



**Figure 10-4 Bank Base Address and Match Structure**

When a match is found on one of the memory banks, its attributes are selected for the functional operation of the external memory access:

- Read-only or read/write operation
- Number of wait states for a single memory access, and for any beat in a burst access
- Burst-inhibit indication. Internal burst requests are still possible during burst-inhibited cycles; the memory controller emulates the burst cycles
- Port size of the external device

Note that if more than one region matches the internal address supplied, then the lowest region is selected to provide the attributes and the chip select.

### 10.2.1 Associated Registers

Status bits for each memory bank are found in the memory control status register (MSTAT). The MSTAT reports write-protect violations for all the banks.

Each of the four banks has a base register (BR) and an option register (OR). The BR<sub>x</sub> and OR<sub>x</sub> registers contain the attributes specific to bank x. The base register contains a valid bit (V) that indicates that the register information for that chip select is valid.

### 10.2.2 Port Size Configuration

The memory controller supports dynamic bus sizing. Defined 8-bit ports can be accessed as odd or even bytes. Defined 16-bit ports, when connected to data bus lines zero to 15, can be accessed as odd bytes, even bytes, or even half-words. Defined 32-bit ports can be accessed as odd bytes, even bytes, odd half-words, even half-words,

or words on word boundaries. The port size is specified by the PS bits in the base register.



### 10.2.3 Write-Protect Configuration

The WP bit in each base register can restrict write access to its range of addresses. Any attempt to write this area results in the associated WPER bit being set in the MSTAT.

If an attempt to access an external device results in a write-protect violation, the memory controller considers the access to be no match. No chip-select line is asserted externally, and the memory controller does not terminate the cycle. The external bus interface generates a normal cycle on the external bus. Since the memory controller does not acknowledge the cycle internally, the cycle may be terminated by external logic asserting  $\overline{TA}$  or by the on-chip bus monitor asserting  $\overline{TEA}$ .

### 10.2.4 Address and Address Space Checking

The base address is written to the BR. The address mask bits for the address are written to the OR. The address type access value, if desired, is written to the AT bits in the BR. The ATM bits in the OR can be used to mask this value. If address type checking is not desired, program the ATM bits to zero.

Each time an external bus cycle access is requested, the address and address type are compared with each one of the banks. If a match is found, the attributes defined for this bank in its BR and OR are used to control the memory access. If a match is found in more than one bank, the lowest bank matched handles the memory access (e.g., bank zero is selected over bank one). Note that when an external master accesses a slave on the bus, the internal AT[0:2] lines reaching the memory controller are forced to 100.

### 10.2.5 Burst Support

Burst support is for read only. The memory controller supports burst accesses of external burstable memory. To enable bursts, clear the  $\overline{BI}$  in the appropriate base register.

Bursts are four beats and non-wrapping. That is, the memory controller executes up to four one-word accesses, but when a modulo four limit is reached, the burst is terminated (even if fewer than four words have been accessed).

When the SIU initiates a burst access, if no match is found in any of the memory controller's regions then a burst access is initiated to the external bus. The termination of each beat for this access is externally controlled (i.e., the user is responsible for terminating each data beat using the bus termination protocol).

To support different types of memory devices, the memory controller supports two types of timing for the BDIP signal: normal and late. Note that the BDIP pin itself is controlled by the external bus interface logic. Refer to [Figure 9-13](#) and [Figure 9-14](#) in [SECTION 9 EXTERNAL BUS INTERFACE](#).

If the memory controller is used to support an external master accessing an external device with bursts, the  $\overline{\text{BDIP}}$  input pin is used to indicate to the memory controller when the burst is terminated.



For addition details, refer to **9.5.3 Burst Transfer**.

### 10.3 Chip-Select Timing

The GPCM allows a glueless and flexible interface between the MPC555 / MPC556 and SRAM, EPROM, EEPROM, ROM devices and external peripherals. When an address and address type matches the values programmed in the BR and OR for one of the memory controller banks, the attributes for the memory cycle are taken from the OR and BR registers as well. These attributes include the following fields: CSNT, ACS, SCY, BSCY, WP, TRLX, BI, PS, and SETA.

Byte write and read-enable signals ( $\overline{\text{WE}}/\overline{\text{BE}}[0:3]$ ) are available for each byte that is written to or read from memory. An output enable ( $\overline{\text{OE}}$ ) signal is provided to eliminate external glue logic for read cycles. Upon system reset, a global (boot) chip select is available. This provides a boot ROM chip select before the system is fully configured. **Table 10-1** summarizes the chip-select timing options.

**Table 10-1 Timing Attributes Summary**

Timing Attribute	Bits/Fields	Description
Access speed	TRLX	The TRLX (timing relaxed) bit determines strobe timing to be fast or relaxed.
Intercycle space time	EHTR	The EHTR (extended hold time on read accesses) bit is provided for devices that have long disconnect times from the data bus on read accesses. EHTR specifies whether the next cycle is delayed one clock cycle following a read cycle, to avoid data bus contentions. EHTR applies to all cycles following a read cycle except for another read cycle to the same region.
Synchronous or asynchronous device	ACS, CSNT	The ACS (address-to-chip-select setup) and CSNT (chip-select negation time) bits cause the timing of the strobes to be the same as the address bus timing, or cause the strobes to have setup and hold times relative to the address bus.
Wait states	SCY, BSCY, SETA, TRLX	From zero to 15 wait states can be programmed for any cycle that the memory controller generates. The transfer is then terminated internally. In simplest case, the cycle length equals $(2 + \text{SCY})$ clock cycles, where SCY represents the programmed number of wait states (cycle length in clocks). The number of wait states is doubled if the TRLX bit is set.  When the SETA (external transfer acknowledge) bit is set, $\overline{\text{TA}}$ must be generated externally, so that external hardware determines the number of wait states.

Note that when a bank is configured for  $\overline{\text{TA}}$  to be generated externally (SETA bit is set) and the TRLX is set, the memory controller requires the external device to provide at least one wait state before asserting  $\overline{\text{TA}}$  to complete the transfer. In this case, the minimum transfer time is three clock cycles.

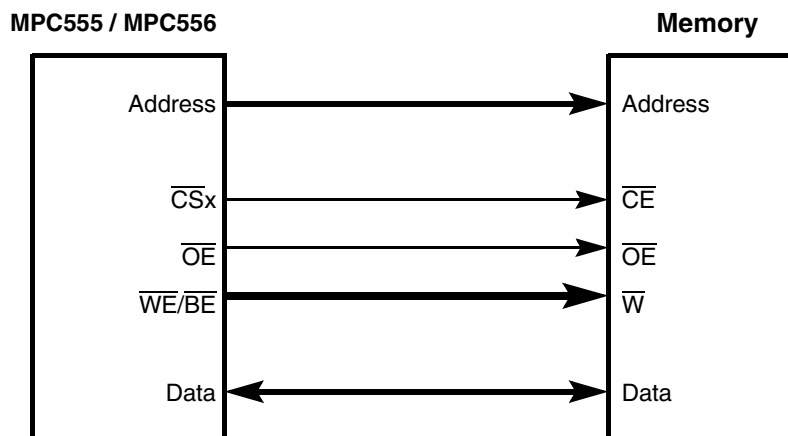
The internal  $\overline{\text{TA}}$  generation mode is enabled if the SETA bit in the OR register is negated. However, if the  $\overline{\text{TA}}$  pin is asserted externally at least two clock cycles before the

wait states counter has expired, this assertion terminates the memory cycle. When SETA is cleared, it is forbidden to assert external TA less than two clocks before the wait states counter expires.



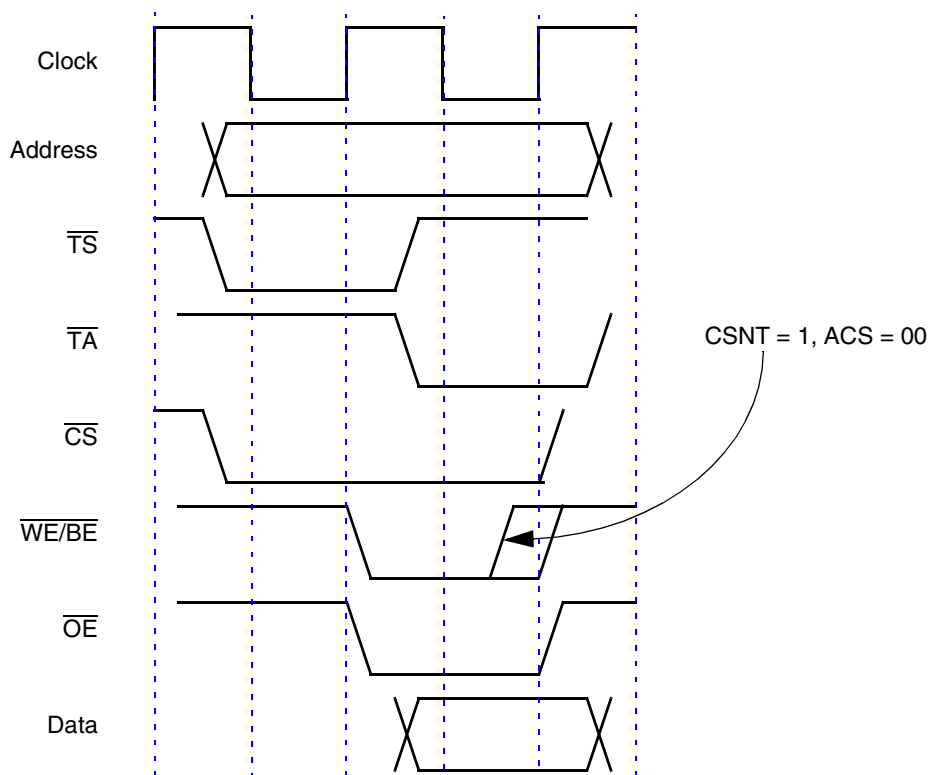
### 10.3.1 Memory Devices Interface Example

**Figure 10-5** describes the basic connection between the MPC555 / MPC556 and a static memory device. In this case  $\overline{CS}_x$  is connected directly to the chip enable ( $\overline{CE}$ ) of the memory device. The  $\overline{WE}/\overline{BE}[0:3]$  lines are connected to the respective  $\overline{W}$  in the memory device where each  $\overline{WE}/\overline{BE}$  line corresponds to a different data byte.



**Figure 10-5 MPC555 / MPC556 GPCM-Memory Devices Interface**

In **Figure 10-6**, the  $\overline{CS}_x$  timing is the same as that of the address lines output. The strobes for the transaction are supplied by the  $\overline{OE}$  and the  $\overline{WE}/\overline{BE}$  lines (if programmed as  $\overline{WE}/\overline{BE}$ ). Because the ACS bits in the corresponding ORx register = 00,  $\overline{CS}$  is asserted at the same time that the address lines are valid. Note that because CSNT is set, the  $\overline{WE}$  signal is negated a quarter of a clock earlier than normal.



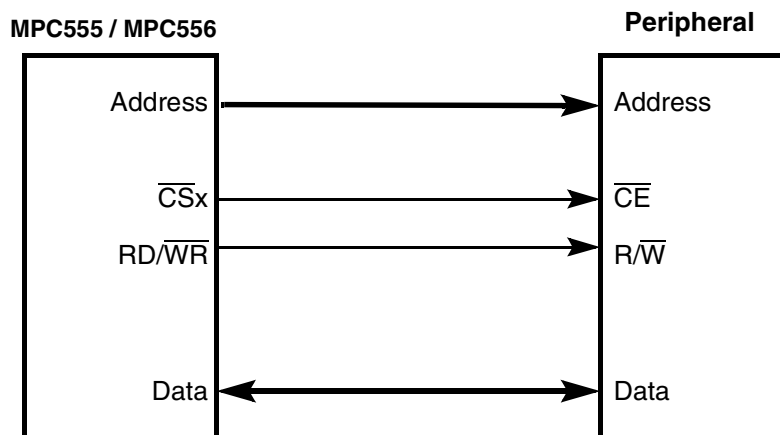
Note: In this and subsequent timing diagrams in this section, the data bus refers to a read cycle. In a write cycle, the data immediately follows  $\overline{TS}$ .

**Figure 10-6 Memory Devices Interface Basic Timing  
(ACS = 00, TRLX = 0)**

### 10.3.2 Peripheral Devices Interface Example

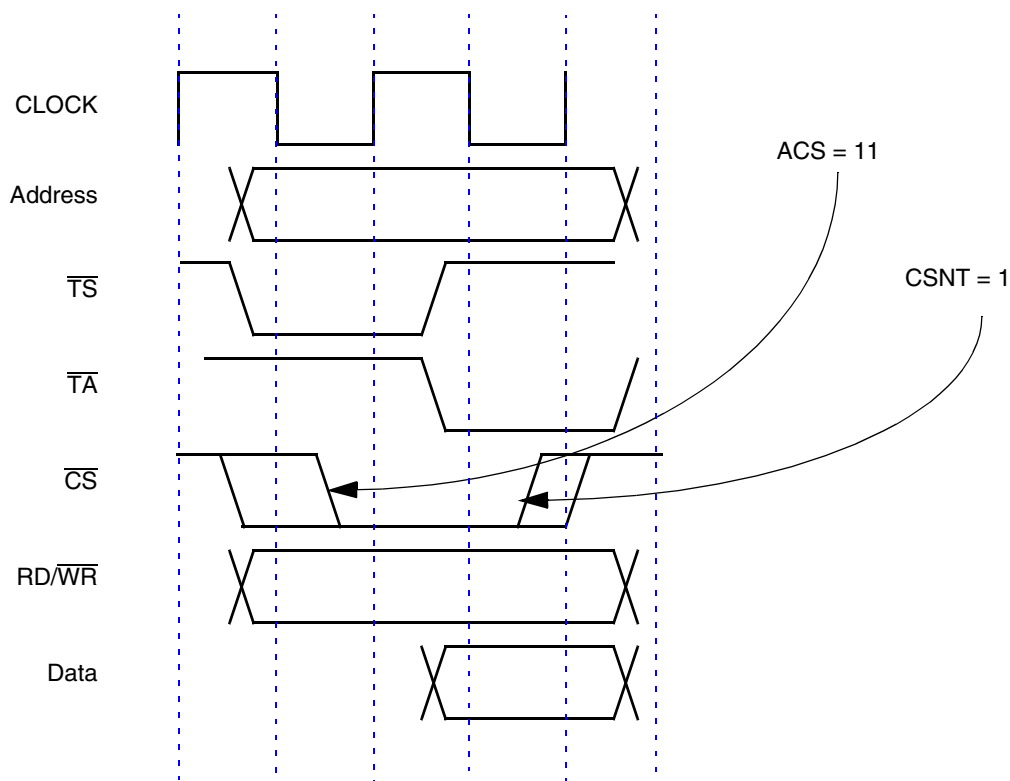
**Figure 10-7** illustrates the basic connection between the MPC555 / MPC556 and an external peripheral device. In this case  $\overline{CSx}$  is connected directly to the chip enable ( $\overline{CE}$ ) of the memory device and the  $R/\overline{W}$  line is connected to the  $R/\overline{W}$  in the peripheral device. The  $\overline{CSx}$  line is the strobe output for the memory access.





**Figure 10-7 Peripheral Devices Interface**

The  $\overline{CS}_x$  timing is defined by the setup time required between the address lines and the  $\overline{CE}$  line. The memory controller allows the user to specify the  $\overline{CS}$  timing to meet the setup time required by the peripheral device. This is accomplished through the ACS field in the base register. In [Figure 10-8](#), the ACS bits are set to 11, so  $\overline{CS}_x$  is asserted half a clock cycle after the address lines are valid.



**Figure 10-8 Peripheral Devices Basic Timing  
(ACS = 11, TRLX = 0)**

### 10.3.3 Relaxed Timing Examples



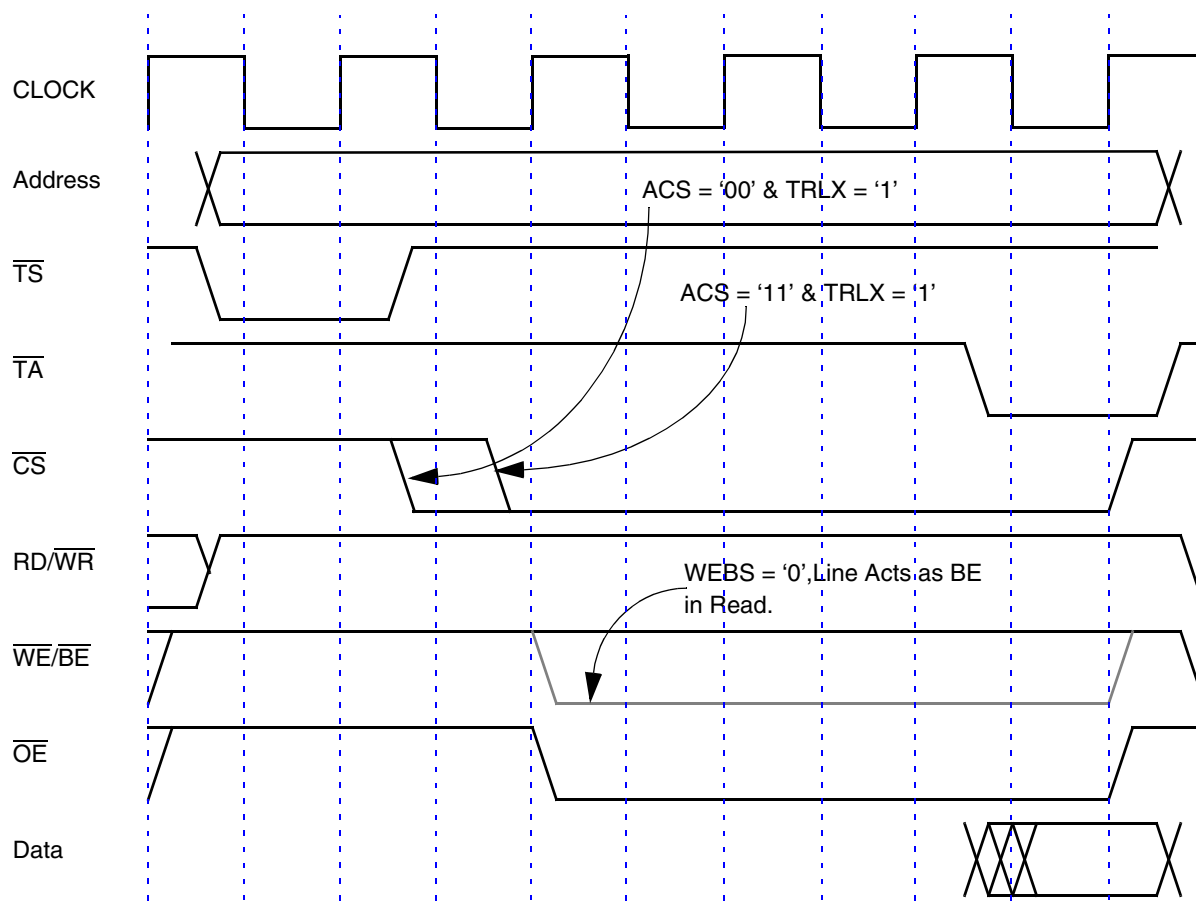
The TRLX field is provided for memory systems that need a more relaxed timing between signals. When TRLX is set and ACS = 0b00, the memory controller inserts an additional cycle between address and strobes ( $\overline{CS}$  line and  $\overline{WE/OE}$ ).

When TRLX and CSNT are both set in a write to memory, the strobe lines ( $\overline{WE/BE}[0:3]$  and  $\overline{CS}$ , if ACS = 0b00) are negated one clock earlier than in the regular case.

Note that in the case of a bank selected to work with external transfer acknowledge (SETA = 1) and TRLX = 1, the memory controller does not support external devices providing  $\overline{TA}$  to complete the transfer with zero wait states. The minimum access duration in this case equals three clock cycles.

**Figure 10-9** shows a read access with relaxed timing. Note the following:

- Strobes ( $\overline{OE}$  and  $\overline{CS}$ ) assertion time is delayed one clock relative to address (TRLX bit set effect).
- Strobe ( $\overline{CS}$ ) is further delayed (half-clock) relative to address due to ACS field being set to 11.
- Total cycle length = 5, is determined as follows:
  - Two clocks for basic cycle
  - SCY = 1 determines 1 wait state, which is multiplied by two due to TRLX being set.
  - Extra clock is added due to TRLX effect on the strobes.

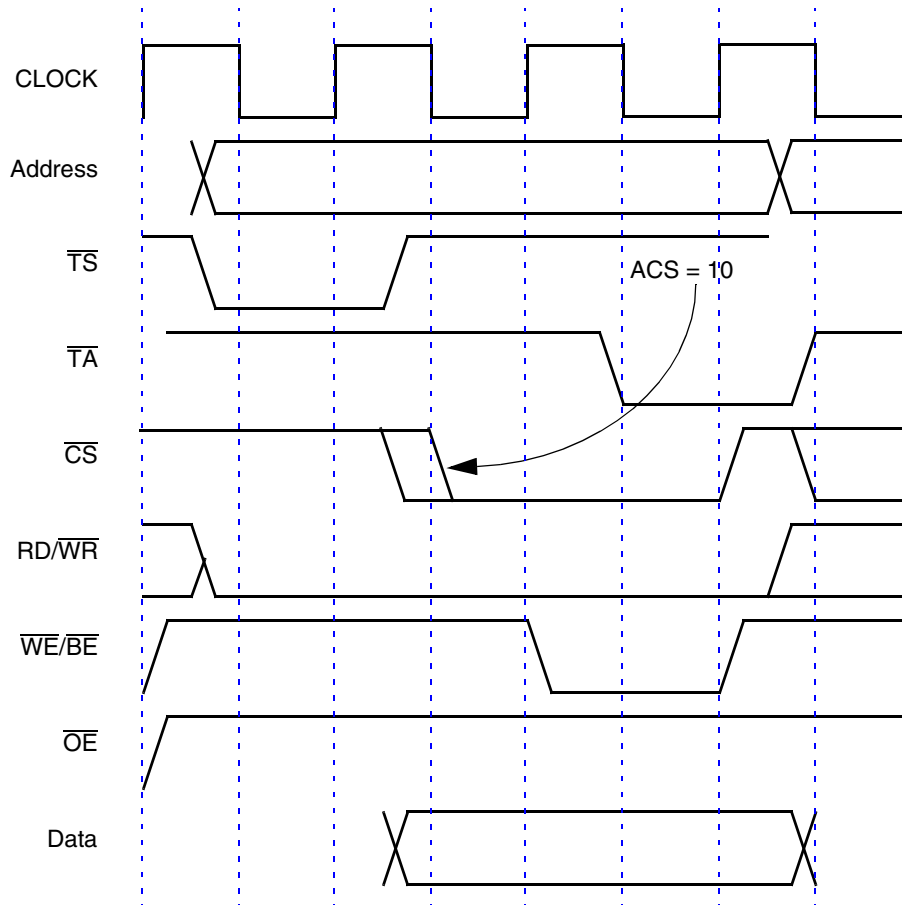


**Figure 10-9 Relaxed Timing–Read Access**  
(ACS = 11, SCY = 1, TRLX = 1)

Figure 10-10 through Figure 10-12 are examples of write accesses using relaxed timing. In Figure 10-10, note the following points:



- Because TRLX is set, assertion of the  $\overline{CS}$  and  $\overline{WE}$  strobes is delayed by one clock cycle.
- $\overline{CS}$  assertion is delayed an additional one quarter clock cycle because  $ACS = 10$ .
- The total cycle length = three clock cycles, determined as follows:
  - The basic memory cycle requires two clock cycles.
  - An extra clock cycle is required due to the effect of TRLX on the strobes.

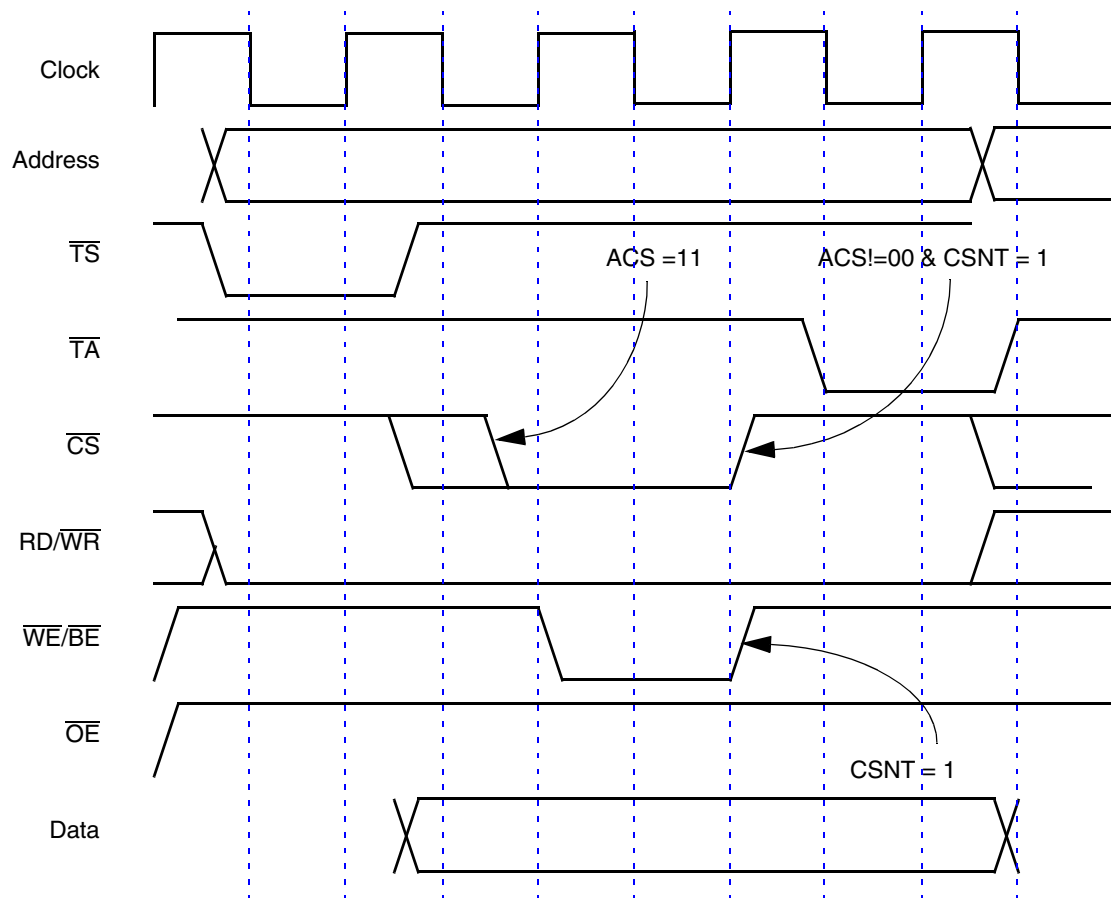


**Figure 10-10 Relaxed Timing–Write Access**  
( $ACS = 10$ ,  $SCY = 0$ ,  $CSNT = 0$ ,  $TRLX = 1$ )

In **Figure 10-11**, note the following:



- Because the TRLX bit is set, the assertion of the  $\overline{CS}$  and  $\overline{WE}$  strobes is delayed by one clock cycle.
- Because ACS = 11, the assertion of CS is delayed an additional half clock cycle.
- Because CSNT = 1,  $\overline{WE}$  is negated one clock cycle earlier than normal. (Refer to **Figure 10-6**). The total cycle length is four clock cycles, determined as follows:
  - The basic memory cycle requires two clock cycles.
  - Two extra clock cycles are required due to the effect of TRLX on the assertion and negation of the  $\overline{CS}$  and  $\overline{WE}$  strobes.

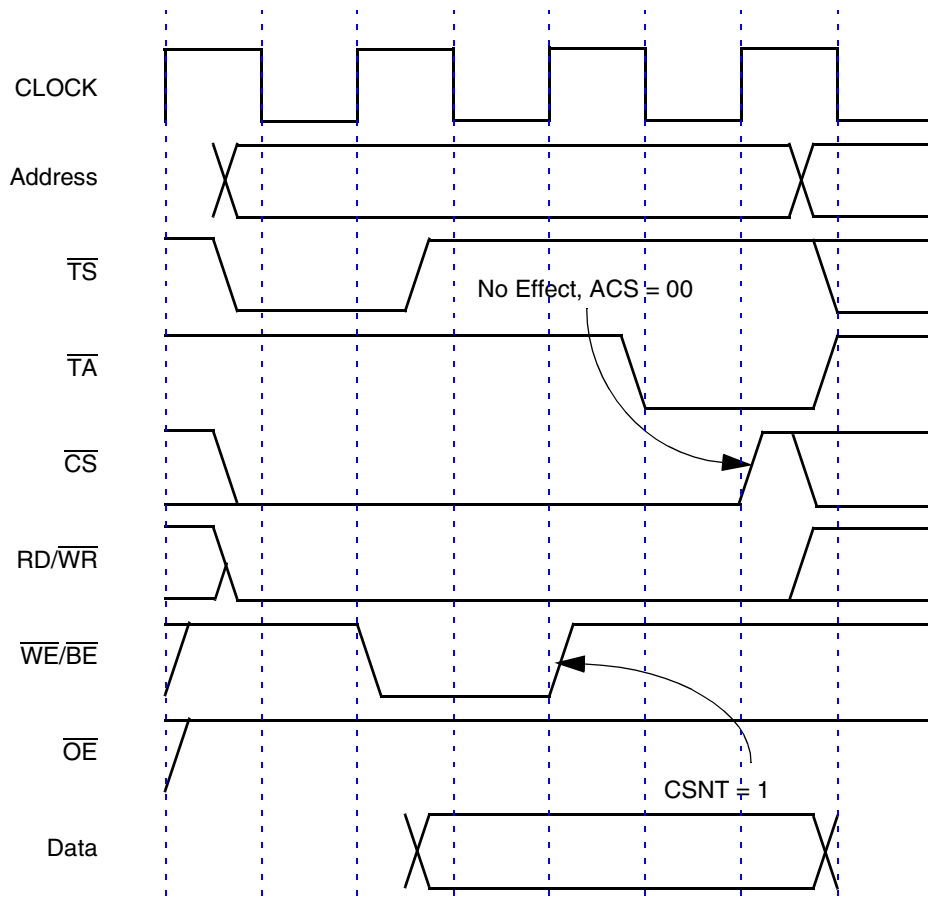


**Figure 10-11 Relaxed Timing – Write Access**  
(ACS = 11, SCY = 0, CSNT = 1, TRLX = 1)

In **Figure 10-12**, notice the following:



- Because  $ACS = 0$ ,  $\overline{TRLX}$  being set does not delay the assertion of the  $\overline{CS}$  and  $\overline{WE}$  strobes.
- Because  $CSNT = 1$ ,  $\overline{WE}/\overline{BE}$  is negated one clock cycle earlier than normal. (Refer to **Figure 10-6**).
- $\overline{CS}$  is not negated one clock cycle earlier, since  $ACS = 00$ .
- The total cycle length is three clock cycles, determined as follows:
  - The basic memory cycle requires two clock cycles.
  - One extra clock cycle is required due to the effect of  $\overline{TRLX}$  on the negation of the  $\overline{WE}/\overline{BE}$  strobes.



**Figure 10-12 Relaxed Timing – Write Access**  
( $ACS = 00$ ,  $SCY = 0$ ,  $CSNT = 1$ ,  $TRLX = 1$ )

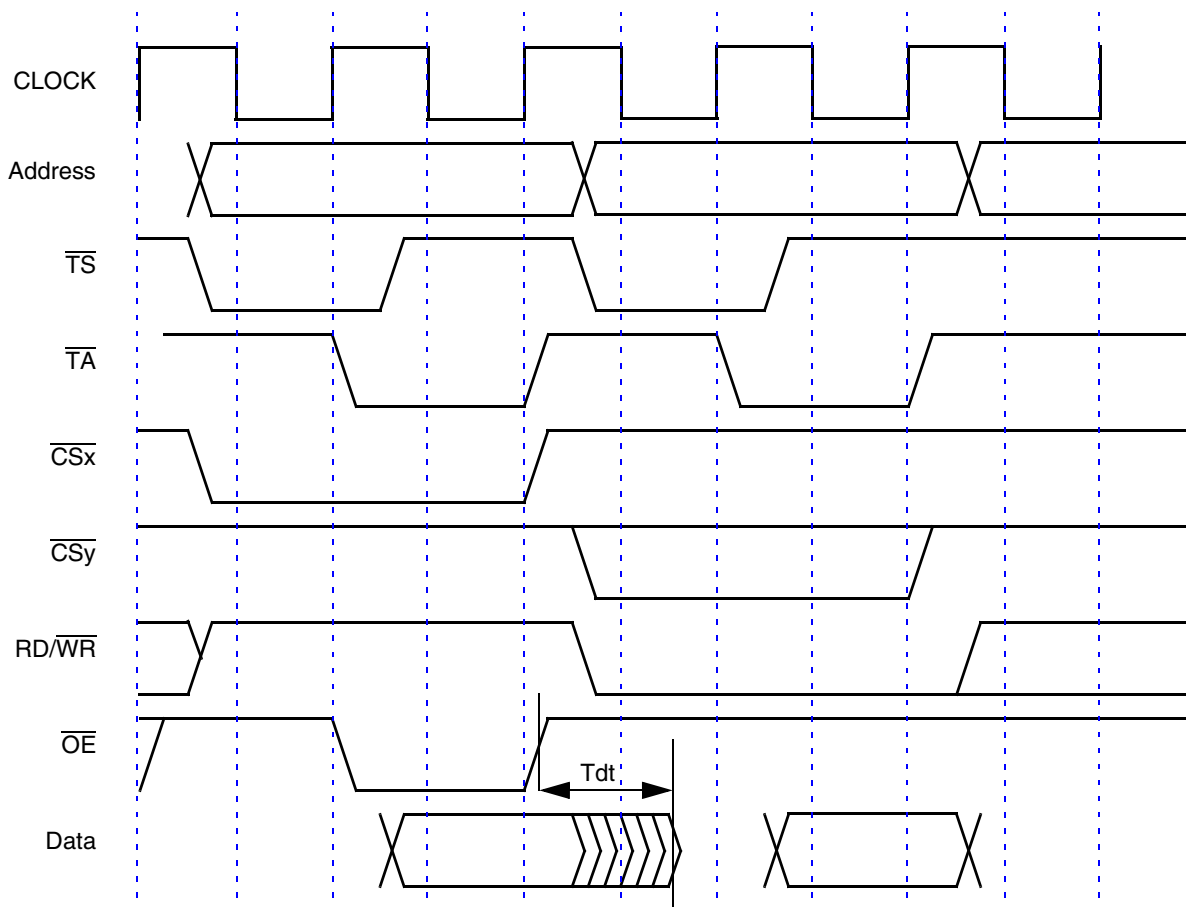
#### 10.3.4 Extended Hold Time on Read Accesses

For devices that require a long disconnection time from the data bus on read accesses, the bit EHTR in the corresponding OR register can be set. In this case any MPC555 / MPC556 access to the external bus following a read access to the referred memory bank is delayed by one clock cycle unless it is a read access to the same bank. **Figure**

10-13 through Figure 10-16 show the effect of the EHTR bit on memory controller timing.

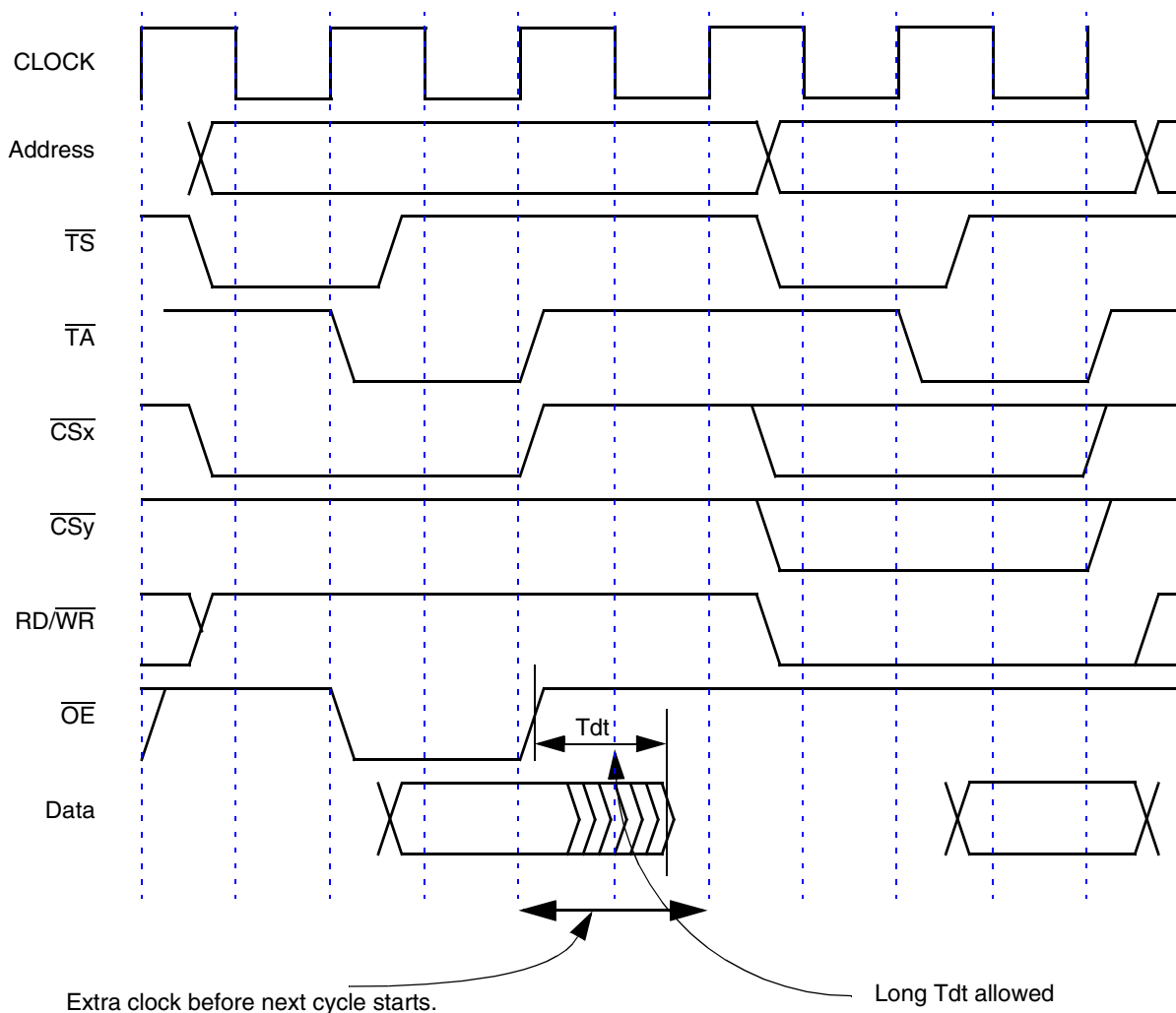


Figure 10-13 shows a write access following a read access. Because EHTR = 0, no extra clock cycle is inserted between memory cycles.



**Figure 10-13 Consecutive Accesses (Write After Read, EHTR = 0)**

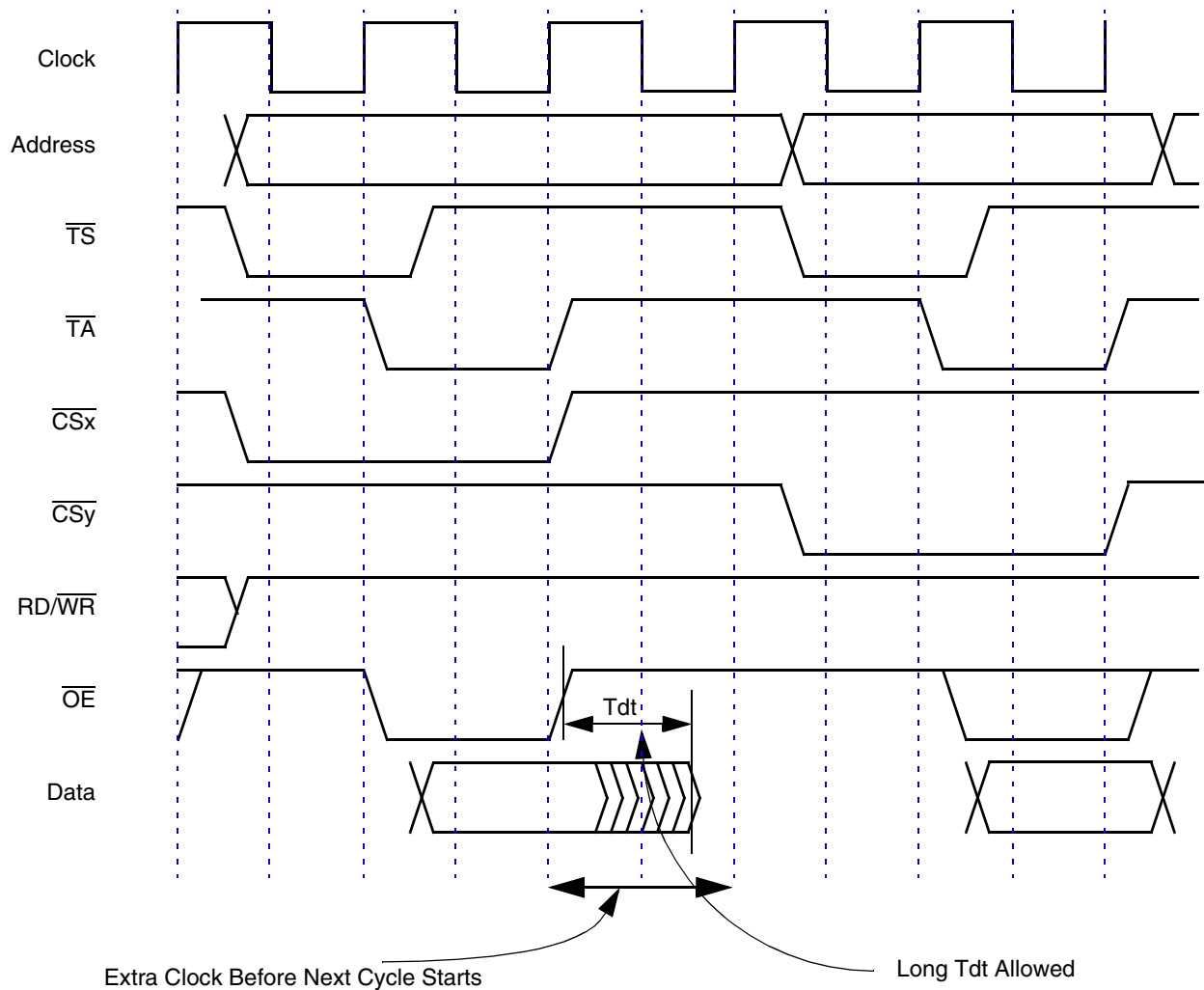
**Figure 10-14** shows a write access following a read access when EHTR = 1. An extra clock is inserted between the cycles. For a write cycle following a read, this is true regardless of whether both accesses are to the same region.



**Figure 10-14 Consecutive Accesses (Write After Read, EHTR = 1)**

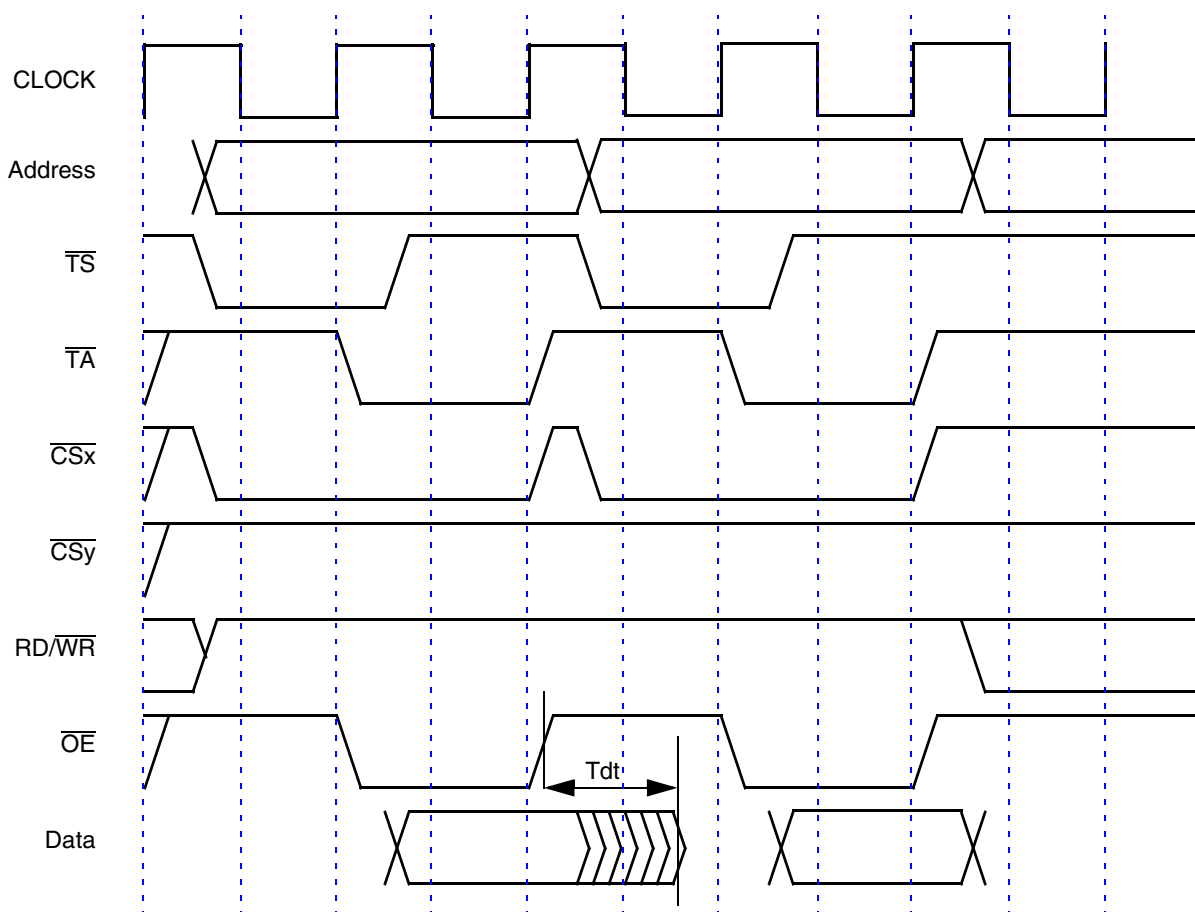


**Figure 10-15** shows consecutive accesses from different banks. Because EHTR = 1 (and the accesses are to different banks), an extra clock cycle is inserted.



**Figure 10-15 Consecutive Accesses  
(Read After Read From Different Banks, EHTR = 1)**

**Figure 10-16** shows two consecutive read cycles from the same bank. Even though  $\text{EHTR} = 1$ , no extra clock cycle is inserted between the memory cycles. (In the case of two consecutive read cycles to the same region, data contention is not a concern.)



**Figure 10-16 Consecutive Accesses**  
(Read After Read From Same Bank,  $\text{EHTR} = 1$ )

### 10.3.5 Summary of GPCM Timing Options

**Table 10-2** summarizes the different combinations of timing options.



**Table 10-2 Programming Rules for Strobes Timing**

TRLX	Access Type	ACS	CSNT	Address to CS Asserted	$\overline{CS}$ Negated to Add/Data Invalid	Address to WE/BE or $\overline{OE}$ Asserted	WE/BE Negated to Add/Data Invalid	$\overline{OE}$ Negated to Add/Data Invalid	Total Number of Cycles
0	Read	00	X	0	$1/4 * \text{clock}$	$3/4 * \text{clock}$	X	$1/4 * \text{clock}$	$2 + \text{SCY}$
0	Read	10	X	$1/4 * \text{clock}$	$1/4 * \text{clock}$	$3/4 * \text{clock}$	X	$1/4 * \text{clock}$	$2 + \text{SCY}$
0	Read	11	X	$1/2 * \text{clock}$	$1/4 * \text{clock}$	$3/4 * \text{clock}$	X	$1/4 * \text{clock}$	$2 + \text{SCY}$
0	Write	00	0	0	$1/4 * \text{clock}$	$3/4 * \text{clock}$	$1/4 * \text{clock}$	X	$2 + \text{SCY}$
0	Write	10	0	$1/4 * \text{clock}$	$1/4 * \text{clock}$	$3/4 * \text{clock}$	$1/4 * \text{clock}$	X	$2 + \text{SCY}$
0	Write	11	0	$1/2 * \text{clock}$	$1/4 * \text{clock}$	$3/4 * \text{clock}$	$1/4 * \text{clock}$	X	$2 + \text{SCY}$
0	Write	00	1	0	$1/4 * \text{clock}$	$3/4 * \text{clock}$	$1/2 * \text{clock}$	X	$2 + \text{SCY}$
0	Write	10	1	$1/4 * \text{clock}$	$1/2 * \text{clock}$	$3/4 * \text{clock}$	$1/2 * \text{clock}$	X	$2 + \text{SCY}$
0	Write	11	1	$1/2 * \text{clock}$	$1/2 * \text{clock}$	$3/4 * \text{clock}$	$1/2 * \text{clock}$	X	$2 + \text{SCY}$
1	Read	00	X	0	$1/4 * \text{clock}$	$3/4 \text{ clock}$	X	$1/4 * \text{clock}$	$2 + 2 * \text{SCY}$
1	Read	10	X	$(1 + 1/4) * \text{clock}$	$1/4 * \text{clock}$	$(1 + 3/4) * \text{clock}$	X	$1/4 * \text{clock}$	$3 + 2 * \text{SCY}$
1	Read	11	X	$(1 + 1/2) * \text{clock}$	$1/4 * \text{clock}$	$(1 + 3/4) * \text{clock}$	X	$1/4 * \text{clock}$	$3 + 2 * \text{SCY}$
1	Write	00	0	0	$1/4 * \text{clock}$	$3/4 \text{ clock}$	$1/4 * \text{clock}$	X	$2 + 2 * \text{SCY}$
1	Write	10	0	$(1 + 1/4) * \text{clock}$	$1/4 * \text{clock}$	$(1 + 3/4) * \text{clock}$	$1/4 * \text{clock}$	X	$3 + 2 * \text{SCY}$
1	Write	11	0	$(1 + 1/2) * \text{clock}$	$1/4 * \text{clock}$	$(1 + 3/4) \text{ clock}$	$1/4 * \text{clock}$	X	$3 + 2 * \text{SCY}$
1	Write	00	1	0	$1/4 * \text{clock}$	$3/4 \text{ clock}$	$(1 + 1/2) * \text{clock}$	X	$3 + 2 * \text{SCY}$
1	Write	10	1	$(1 + 1/4) * \text{clock}$	$(1 + 1/2) * \text{clock}$	$(1 + 3/4) \text{ clock}$	$(1 + 1/2) * \text{clock}$	X	$4 + 2 * \text{SCY}$
1	Write	11	1	$(1 + 1/2) * \text{clock}$	$(1 + 1/2) * \text{clock}$	$(1 + 3/4) \text{ clock}$	$(1 + 1/2) * \text{clock}$	X	$4 + 2 * \text{SCY}$

NOTE: Timing in this table refers to the typical timing only. Consult the electrical characteristics for exact worst-case timing values.  $1/4$  clock actually means 0 to  $1/4$  clock,  $1/2$  clock means  $1/4$  to  $1/2$  clock.

Additional timing rules not covered in [Table 10-2](#) include the following:

- If  $\text{SETA} = 1$ , an external  $\overline{TA}$  signal is required to terminate the cycle.
- If  $\text{TRLX} = 1$  and  $\text{SETA} = 1$ , the minimum cycle length = 3 clock cycles (even if  $\text{SCY} = 0000$ )
- If  $\text{TRLX} = 1$ , the number of wait states =  $2 * \text{SCY} \& 2 * \text{BSCY}$
- If  $\text{EHTR} = 1$ , an extra (idle) clock cycle is inserted between a read cycle and a following read cycle to another region, or between a read cycle and a following write cycle to any region.
- If  $\text{LBDIP} = 1$  (late  $\overline{BDIP}$  assertion), the  $\overline{BDIP}$  pin is asserted only after the number of wait states for the first beat in a burst have elapsed. See [Figure 9-13](#) in [SECTION 9 EXTERNAL BUS INTERFACE](#) as well as [9.5.4 Burst Mechanism](#). Note that this function can operate only when the cycle termination is internal, using the number of wait states programmed in one of the ORx registers

## 10.4 Global (Boot) Chip-Select Operation

Global (boot) chip-select operation allows address decoding for a boot ROM before system initialization.  $\overline{CS}[0]$  is the global chip-select output. Its operation differs from that of the other external chip-select outputs following a system reset. When the RCPU begins accessing memory after a system reset,  $\overline{CS}[0]$  is asserted for every address, unless an internal device (register) is accessed.

The global chip select provides a programmable port size at system reset using the reset BPS pins ([4:5]) of the reset configuration word, allowing a boot ROM to be located anywhere in the address space. For more information, see [7.5.2 Hard Reset Configuration Word](#). The global chip select does not provide write protection and responds to all address types.  $\overline{CS}[0]$  operates in this way until the first write to the  $\overline{CS}[0]$  option register (OR0). The pin can be programmed to continue decoding a range of addresses after this write, provided the preferred address range is first loaded into base register zero. After the first write to OR0, the global chip select can only be restarted with a system reset.

The memory controller operates in boot mode until the user modifies the values in OR0 and BR to the ones desired.

**Table 10-3** shows the initial values of the “boot bank” in the memory controller.

**Table 10-3 Boot Bank Fields Values After Hard Reset**

Field	Value (Binary)
PS	From reset configuration
WP	0
V	From reset configuration
AM[0:16]	0 0000 0000 0000 0000
ATM[0:2]	000
CSNT	0
ACS[0:1]	00
BI	1
SCY[0:3]	1111
BSCY[0:2]	011
SETA	0
TRLX	0

### NOTE

If the MPC555 / MPC556 is configured (in the reset configuration word) to use the internal flash EEPROM as boot memory  $\overline{CS}[0]$  is not asserted.

## 10.5 Write and Byte Enable Signals

The GPCM determines the timing and value of the  $\overline{WE}/\overline{BE}$  signals if allowed by the port size of the accessed bank, the transfer size of the transaction and the address accessed.

The functionality of the  $\overline{WE}/\overline{BE}[0:3]$  pins depends upon the value of the write enable/byte select (WEBS) bit in the corresponding BR register. Setting WEBS to 1 will enable these pins as  $\overline{BE}$ , while resetting it to zero will enable them as  $\overline{WE}$ .  $\overline{WE}$  is asserted only during write access, while  $\overline{BE}$  is asserted for both read and write accesses. The timing of the  $\overline{WE}/\overline{BE}$  pins remains the same in either case, and is determined by the TRLX, ACS and CSNT bits.

The upper  $\overline{WE}/\overline{BE}$  ( $\overline{WE}[0]/\overline{BE}[0]$ ) indicates that the upper eight bits of the data bus (D0–D7) contains valid data during a write/read cycle. The upper-middle write byte enable ( $\overline{WE}[1]/\overline{BE}[1]$ ) indicates that the upper-middle eight bits of the data bus (D8–D15) contains valid data during a write/read cycle. The lower-middle write byte enable ( $\overline{WE}[2]/\overline{BE}[2]$ ) indicates that the lower-middle eight bits of the data bus (D16–D23) contains valid data during a write/read cycle. The lower write/read enable ( $\overline{WE}[3]/\overline{BE}[3]$ ) indicates that the lower eight bits of the data bus contains valid data during a write cycle.

The write/byte enable lines affected in a transaction for 32-bit port (PS = 00), a 16-bit port (PS = 10) and a 8-bit port (PS = 01) are shown in [Table 10-4](#). This table shows which write enables are asserted (indicated with an 'X') for different combinations of port size and transfer size

**Table 10-4 Write Enable/Byte Enable Signals Function**

Transfer Size	TSIZ		Address		32-bit Port Size				16-bit Port Size				8-bit Port Size			
			A30	A31	$\overline{WE}[0]$ $\overline{BE}[0]$	$\overline{WE}[1]$ $\overline{BE}[1]$	$\overline{WE}[2]$ $\overline{BE}[2]$	$\overline{WE}[3]$ $\overline{BE}[3]$	$\overline{WE}[0]$ $\overline{BE}[0]$	$\overline{WE}[1]$ $\overline{BE}[1]$	$\overline{WE}[2]$ $\overline{BE}[2]$	$\overline{WE}[3]$ $\overline{BE}[3]$	$\overline{WE}[0]$ $\overline{BE}[0]$	$\overline{WE}[1]$ $\overline{BE}[1]$	$\overline{WE}[2]$ $\overline{BE}[2]$	$\overline{WE}[3]$ $\overline{BE}[3]$
Byte	0	1	0	0	X				X				X			
	0	1	0	1		X				X			X			
	0	1	1	0			X		X				X			
	0	1	1	1				X		X			X			
Half-Word	1	0	0	0	X	X			X	X			X			
	1	0	1	0			X	X	X	X			X			
Word	0	0	0	0	X	X	X	X	X	X			X			

## 10.6 Dual Mapping of the Internal Flash EEPROM Array

The user can enable mapping of the internal flash EEPROM (CMF) module to an external memory region controlled by the memory controller. Only one region can be programmed to be dual-mapped. When dual mapping is enabled (DME bit is set in DMBR), an internal address matches the dual-mapped address range (as programmed in the DMBR), and the cycle type matches AT/ATM field in DMBR/DMOR registers, then the following occur:

- The internal flash memory does not respond to that address



- The memory controller takes control of the external access
- The attributes for the access are taken from one of the base and option registers of the appropriate chip select
- The chip-select region selected is determined by the “ $\overline{CS}$  line select” bit field (**10.8.5 Dual Mapping Base Register (DMBR)**).

Note that dual mapping can operate only for addresses within the FLASH pre-allocated address (up to 2 Mbytes). This is achieved by programming only six bits for the base address (11:16); The upper bits are always set as follows:

$$\text{bus\_addr}[0:10] = \{0000000, \text{isb}[0:2], 0\}$$

Where ISB[0:2] represents the bit field in IMMR register that determines the location of the address map of the MPC555 / MPC556.

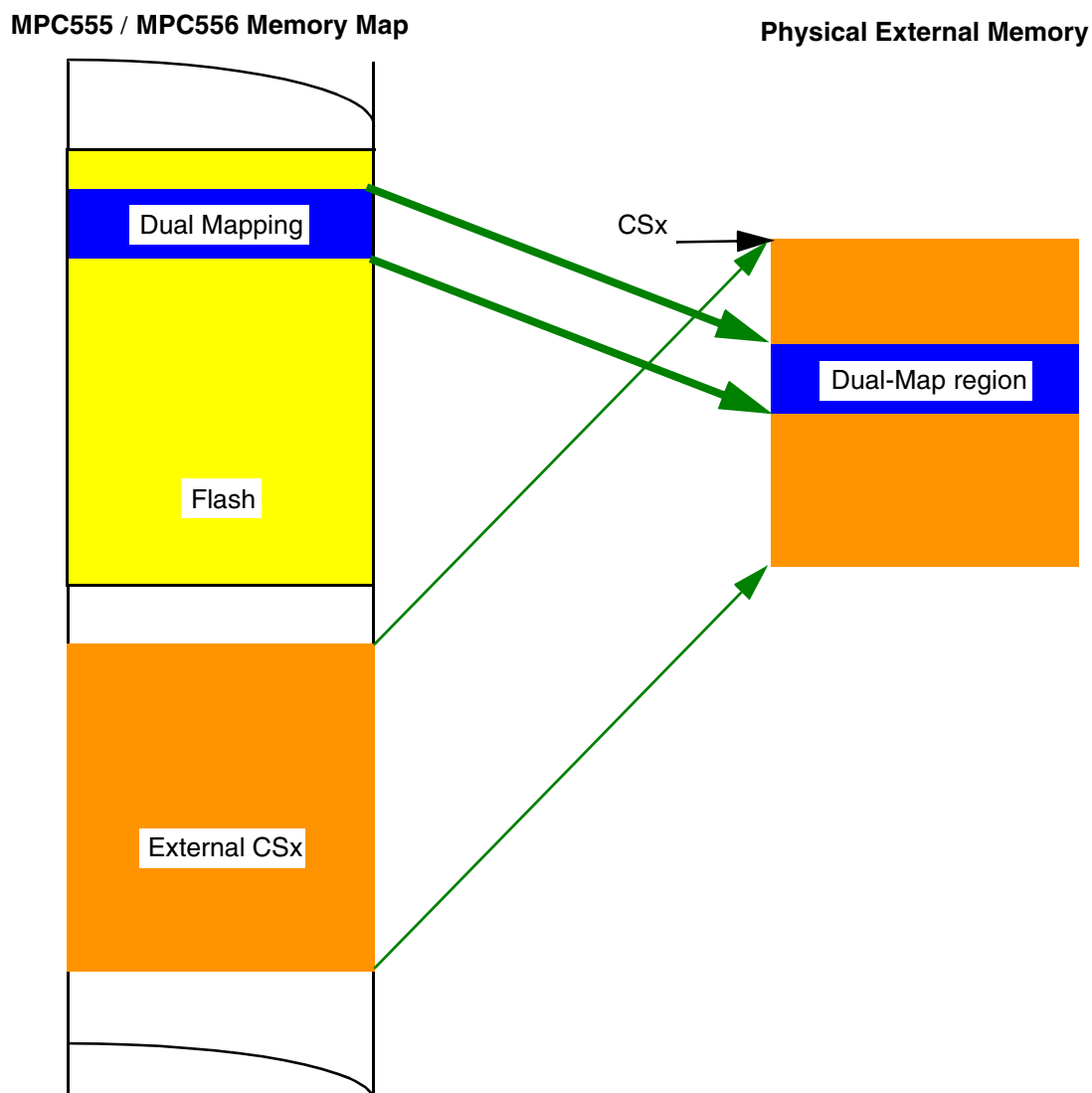
With dual mapping, aliasing of address spaces may occur. This happens when the user maps the dual-mapped region into a region which is also mapped into one of the four regions available in the memory controller. If the user writes code or data to the dual-mapped region, care must be taken to avoid overwriting this code or data by normal accesses of the chip-select region.

There is a match if:

$$\text{bus\_address}[0:16] == \{0000000, \text{isb}[0:2], 0, \text{dmbr\_reg\_value}[1:6]\}$$

Care must also be taken to avoid overwriting “normal”  $\overline{CS}_x$  data with dual-mapped code or data.

One way to avoid this situation is by disabling the chip-select region and enabling only the dual-mapped region (DME = 1, but  $V_x = 0$ , where  $x$  = selected region, 0.3). **Figure 10-17** illustrates the phenomena.



**Figure 10-17 Aliasing Phenomena Illustration**

The default state is to allow dual-mapping data accesses only; this means that dual mapping is possible only for data accesses on the internal bus. Also, the default state takes the lower two Mbytes of the MPC555 / MPC556 internal flash memory. Hence, caution should be taken to change the dual-mapping setup before the first data access.

#### **NOTE**

Dual mapping is *not* supported for an external master when the memory controller serves the access; In such a case, the MPC555 / MPC556 terminates the cycle by asserting  $\overline{\text{TEA}}$ .

## 10.7 Memory Controller External Master Support



The memory controller in the MPC555 / MPC556 supports accesses initiated by both internal and external bus masters to external memories. If the address of any master is mapped within the internal MPC555 / MPC556 address space, the access will be directed to the internal device, and will be ignored by the memory controller. If the address is not mapped internally, but rather mapped to one of the memory controller regions, the memory controller will provide the appropriate chip select and strobes as programmed in the corresponding region (see [6.13.1.3 External Master Control Register \(EMCR\)](#)).

The MPC555 / MPC556 supports only synchronous external bus masters. This means that the external master works with CLKOUT and implements the MPC555 / MPC556 bus protocol to access a slave device.

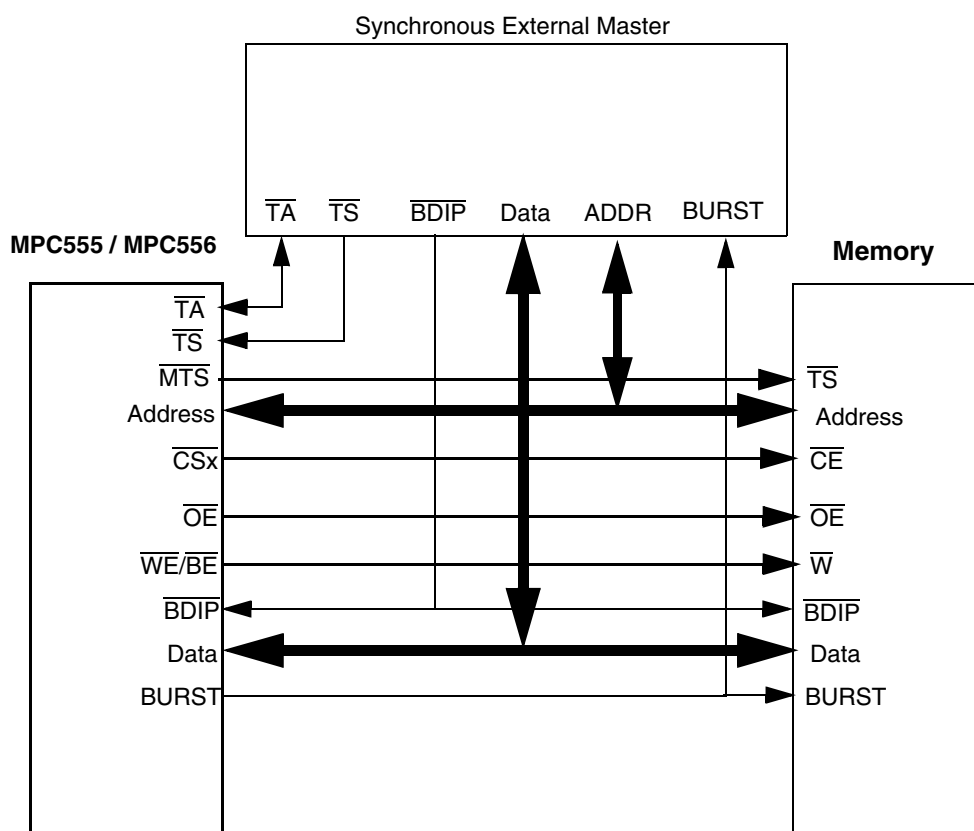
A synchronous master initiates a transfer by asserting  $\overline{TS}$ . The ADDR[0:31] signals must be stable from the rising edge of CLKOUT during which  $\overline{TS}$  is sampled, until the last  $\overline{TA}$  acknowledges the transfer. Since the external master works synchronously with the MPC555 / MPC556, only setup and hold times around the rising edge of CLKOUT are important. Once the  $\overline{TS}$  is detected/asserted, the memory controller compares the address with each one of its defined valid banks to find a possible match. But, since the external address space is shorter than the internal space, the actual address that is used for comparing against the memory controller regions is in the format of: {00000000, bits 8:16 of the external address}. In the case where a match is found, the controls to the memory devices are generated and the transfer acknowledge indication ( $\overline{TA}$ ) is supplied to the master.

Since it takes two clocks for the external address to be recognized and handled by the memory controller, the  $\overline{TS}$  which is generated by the external master is ahead of the corresponding  $\overline{CS}$  and strobes which are asserted by the memory controller. This 2-clock delay might cause problems in some synchronous memories. To overcome this, the memory controller generates the  $\overline{MTS}$  (memory transfer start) strobe which can be used in the slave's memory instead of the external master's  $\overline{TS}$  signal. As seen in [Figure 10-18](#), the  $\overline{MTS}$  strobe is synchronized to the assertion of  $\overline{CS}$  by the memory controller so that the external memory can latch the external master's address correctly. To activate this feature, the MTSC bit must be set in the SIUMCR register. Refer to [6.13.1.1 SIU Module Configuration Register](#) for more information.

When the external master accesses the internal flash when it is disabled, then the access is terminated with transfer error acknowledge ( $\overline{TEA}$  pin) asserted, and the memory controller does not support this access in any way.

When the memory controller serves an external master, the  $\overline{BDIP}$  pin becomes an input pin. This pin is watched by the memory controller to detect when the burst is terminated.





Note that the memory controller's  $\overline{BDIP}$  line is used as a burst\_in\_progress signal.

**Figure 10-18 Synchronous External Master Configuration For GPCM-Handled Memory Devices**



### Figure 10-19 Synchronous External Master Basic Access (GPCM Controlled)

## 10.8 Programming Model

The following registers are used to control the memory controller.



**Table 10-5 Memory Controller Address Map**

Address	Register
0x2F C100	Base Register Bank 0 (BR0)
0x2F C104	Option Register Bank 0 (OR0)
0x2F C108	Base Register Bank 1 (BR1)
0x2F C10C	Option Register Bank 1 (OR1)
0x2F C110	Base Register Bank 2 (BR2)
0x2F C114	Option Register Bank 2 (OR2)
0x2F C118	Base Register Bank 3 (BR3)
0x2F C11C	Option Register Bank 3 (OR3)
0x2F C120 — 0x13F	Reserved
0x2F C140	Dual-Mapping Base Register (DMBR)
0x2F C144	Dual-Mapping Option Register (DMOR)
0x2F C148 — 0x2F C174	Reserved
0x2F C178	Memory Status Register (MSTAT)

Note:

In all subsequent registers bit tables, if two reset values are given: the upper is for  $\overline{CS}_x$ ,  $x = 1, 2, 3$ , and the lower is dedicated to  $\overline{CS}[0]$ .

### 10.8.1 General Memory Controller Programming Notes

1. In the case of an external master that accesses an internal MPC555 / MPC556 module (in slave or peripheral mode), if that slave device address also matches one of the memory controller's regions, the memory controller will not issue any  $\overline{CS}$  for this access, nor will it terminate the cycle. Thus, this practice should be avoided. Be aware also that any internal slave access prevents memory controller operation.
2. If the memory controller serves an external master, then it can support accesses to 32-bit port devices only. This is because the MPC555 / MPC556 external bus interface cannot initiate extra cycles to complete an access to a smaller port-size device as it does not own the external bus.
3. When the SETA bit in the base register is set, then the timing programming for the various strobes ( $\overline{CS}$ ,  $\overline{OE}$  and  $\overline{WE/BE}$ ) may become meaningless.

## 10.8.2 Memory Controller Status Registers (MSTAT)

### MSTAT — Memory Controller Status Register

0x2F C178

MSB	1	2	3	4	5	6	7	8	9	10	11	12	13	14	LSB
0															15
RESERVED								WPER 0	WPER 1	WPER 2	WPER 3	RESERVED			
HARD RESET:															
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0

**Table 10-6 MSTAT Bit Descriptions**

Bit(s)	Name	Description
0:7	—	Reserved
8:11	WPER0 – WPER3	Write protection error for bank x. This bit is asserted when a write-protect error occurs for the associated memory bank. A bus monitor (responding to $\overline{TEA}$ assertion) will, if enabled, prompt the user to read this register if $\overline{TA}$ is not asserted during a write cycle. WPERx is cleared by writing one to the bit or by performing a system reset. Writing a zero has no effect on WPER.
12:15	—	Reserved

## 10.8.3 Memory Controller Base Registers (BR0 – BR3)

### BR0 – BR3 — Memory Controller Base Registers 0 – 3 0x2F C100, C108, C110, C118

MSB 0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
BA															
HRESET															
U	U	U	U	U	U	U	U	U	U	U	U	U	U	U	U
16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	LSB 31
BA	AT			PS		RE-SERV ED	WP	RESERVED		WEBS	TBDIP	LBDIP	SETA	BI	V
HRESET															
U	U	U	U	ID[4:5]*		0	0	0	0	0	0	0		1	$\overline{ID3^{**}}$

\* Reset value is determined by the value on the internal data bus during reset.

\*\* The BR0 Reset value is determined by the value on the internal data bus during reset (reset-configuration word). The reset value of the V bit of BR1-3 = 0.

**Table 10-7 BR0 – BR3 Bit Descriptions**



Bit(s)	Name	Description
0:16	BA	Base address. These bits are compared to the corresponding unmasked address signals among ADDR[0:16] to determine if a memory bank controlled by the memory controller is being accessed by an internal bus master. (The address types are also compared.) These bits are used in conjunction with the AM[0:16] bits in the OR.
17:19	AT	Address type. This field can be used to require accesses of the memory bank to be limited to a certain address space type. These bits are used in conjunction with the ATM bits in the OR. Note that the address type field uses only AT[0:2] and does not need AT[3] to define the memory type space. For a full definition of address types, refer to <a href="#">9.5.7.6 Address Types</a> .
20:21	PS	Port size 00 = 32-bit port 01 = 8-bit port 10 = 16-bit port 11 = Reserved
22	—	Reserved
23	WP	Write protect. An attempt to write to the range of addresses specified in a base address register that has this bit set can cause the $\overline{TEA}$ signal to be asserted by the bus-monitor logic (if enabled), causing termination of this cycle. 0 = Both read and write accesses are allowed 1 = Only read accesses are allowed. The $\overline{CSx}$ signal and $\overline{TA}$ are not asserted by the memory controller on write cycles to this memory bank. WPER is set in the MSTAT register if a write to this memory bank is attempted
24:25	—	Reserved
26	WEBS	Write-enable/byte-select. This bit controls the functionality of the $\overline{WE}/\overline{BE}$ pads. 0 = The $\overline{WE}/\overline{BE}$ pads operate as $\overline{WE}$ 1 = The $\overline{WE}/\overline{BE}$ pads operate as $\overline{BE}$
27	TBDIP	Toggle-burst data in progress. TBDIP determines how long the $\overline{BDIP}$ strobe will be asserted for each data beat in the burst cycles.
28	LBDIP	Late-burst-data-in-progress (LBDIP). This bit determines the timing of the first assertion of the $\overline{BDIP}$ pin in burst cycles. Note: it is not allowed to set both LBDIP and TBDIP bits in a region's base registers; the behavior of the design in such cases is unpredictable. 0 = Normal timing for $\overline{BDIP}$ assertion (asserts one clock after negation of $\overline{TS}$ ) 1 = Late timing for $\overline{BDIP}$ assertion (asserts after the programmed number of wait states)
29	SETA	External transfer acknowledge 0 = $\overline{TA}$ generated internally by memory controller 1 = $\overline{TA}$ generated by external logic. Note that programming the timing of $\overline{CS}/\overline{WE}/\overline{OE}$ strobes may have no meaning when this bit is set
30	BI	Burst inhibit 0 = Memory controller drives $\overline{BI}$ negated (high). The bank supports burst accesses. 1 = Memory controller drives $\overline{BI}$ asserted (low). The bank does not support burst accesses. Note: Following a system reset, the $\overline{BI}$ bit is set in OR0.
31	V	Valid bit. When set, this bit indicates that the contents of the base-register and option-register pair are valid. The $\overline{CS}$ signal does not assert until the V-bit is set. Note that an access to a region that has no V-bit set may cause a bus monitor timeout. Note also that following a system reset, the V-bit in BR0 reflects the value of $\overline{ID3}$ in the reset configuration word.

## 10.8.4 Memory Controller Option Registers (OR0 – OR3)

### OR0 – OR3 — Memory Controller Option Registers 0 – 3

0x2F C104, C10C,  
C114, C11C



MSB 0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	
AM*																
HRESET: (OR[1:3])																
U	U	U	U	U	U	U	U	U	U	U	U	U	U	U	U	
HRESET (OR0)																
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	
16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	LSB 31	
AM*	ATM			CSNT	ACS		EHTR	SCY				BSCY			TRLX	
HRESET: (OR[1:3]):																
U	U	U	U	U	U	U	U	U	U	U	U	U	U	U	U	
HRESET (OR0)																
0	0	0	0	0	0	0	0	1	1	1	1	0	1	1	0	

\*It is recommended that this field would hold values that are the power of 2 minus 1 (e.g.,  $2^3 - 1 = 7$  [0b111]).

**Table 10-8 OR0 – OR3 Bit Descriptions**

Bit(s)	Name	Description
0:16	AM	Address mask. This field allows masking of any corresponding bits in the associated base register. Masking the address bits independently allows external devices of different size address ranges to be used. Any clear bit masks the corresponding address bit. Any set bit causes the corresponding address bit to be used in comparison with the address pins. Address mask bits can be set or cleared in any order in the field, allowing a resource to reside in more than one area of the address map. This field can be read or written at anytime. Following a system reset, the AM bits are reset in OR0.
17:19	ATM	Address type mask. This field masks selected address type bits, allowing more than one address space type to be assigned to a chip-select. Any set bit causes the corresponding address type code bits to be used as part of the address comparison. Any cleared bit masks the corresponding address type code bit. Clear the ATM bits to ignore address type codes as part of the address comparison. Note that the address type field uses only AT[0:2] and does not need AT[3] to define the memory type space. Following a system reset, the ATM bits are reset in OR0.
20	CSNT	Chip-select negation time. Following a system reset, the CSNT bit is reset in OR0. 0 = $\overline{CS/WE}$ are negated normally. 1 = $\overline{CS/WE}$ are negated a quarter of a clock earlier than normal Following a system reset, the CSNT bit is reset in OR0.
21:22	ACS	Address to chip-select setup. Following a system reset, the ACS bits are reset in OR0. 00 = $\overline{CS}$ is asserted at the same time that the address lines are valid. 01 = Reserved 10 = $\overline{CS}$ is asserted a quarter of a clock after the address lines are valid. 11 = $\overline{CS}$ is asserted half a clock after the address lines are valid Following a system reset, the ACS bits are reset in OR0.

**Table 10-8 OR0 – OR3 Bit Descriptions (Continued)**



Bit(s)	Name	Description
23	EHTR	Extended hold time on read accesses. This bit, when asserted, inserts an idle clock cycle after a read access from the current bank and any MPC555 / MPC556 write accesses or read accesses to a different bank. 0 = Memory controller generates normal timing 1 = Memory controller generates extended hold timing
24:27	SCY	Cycle length in clocks. This four-bit value represents the number of wait states inserted in the single cycle, or in the first beat of a burst, when the GPCM handles the external memory access. Values range from 0 (0b0000) to 15 (0b1111). This is the main parameter for determining the length of the cycle. The total cycle length may vary depending on the settings of other timing attributes. The total memory access length is (2 + SCY) x Clocks. If the user has selected an external $\overline{TA}$ response for this memory bank (by setting the SETA bit), then the SCY field is not used. <b>NOTE:</b> Following a system reset, the SCY bits are set to 0b1111 in OR0.
28:30	BSCY	Burst beats length in clocks. This field determines the number of wait states inserted in all burst beats except the first, when the GPCM starts handling the external memory access and thus using SCY[0:3] as the main parameter for determining the length of that cycle. The total cycle length may vary depending on the settings of other timing attributes. The total memory access length for the beat is (1 + BSCY) x Clocks. If the user has selected an external $\overline{TA}$ response for this memory bank (by setting the SETA bit) then BSCY[0:3] are not used. 000 = 0-clock-cycle (1 clock per data beat) 001 = 1-clock-cycle wait states (2 clocks per data beat) 010 = 2-clock-cycle wait states (3 clocks per data beat) 011 = 3-clock-cycle wait states (4 clocks per data beat) 1xx = Reserved
31	TRLX	Timing relaxed. This bit, when set, modifies the timing of the signals that control the memory devices during a memory access to this memory region. Relaxed timing multiplies by two the number of wait states determined by the SCY and BSCY fields. Refer to <a href="#">10.3.5 Summary of GPCM Timing Options</a> for a full list of the effects of this bit on pins timing. 0 = Normal timing is generated by the GPCM. 1 = Relaxed timing is generated by the GPCM Following a system reset, the TRLX bit is set in OR0.

### 10.8.5 Dual Mapping Base Register (DMBR)

#### DMBR — Dual Mapping Base Register

**0x2F C140**

MSB 0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
0	BA					RESERVED			AT			RESERVED			
HARD RESET:															
0	U	U	U	U	U	U	0	0	0	0	0	1	0	0	0
16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	LSB 31
RESERVED												DMCS		DME	
HARD RESET:															
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	ID31*

\*The reset value is a reset configuration word value extracted from the indicated internal data bus lines.

**Table 10-9 DMBR Bit Descriptions**



Bit(s)	Name	Description
0	—	Reserved
1:6	BA	Base address. The base address field is compared (along with the address type field) to the address of the address bus to determine whether an address should be dual-mapped by one of the memory banks controlled by the memory controller. These bits are used in conjunction with the AM[11:16] bits in the OR.  Bit 10: is cleared at reset. That way, the default range for the dual mapping is 2 Mbytes. Note that by setting this bit, the range becomes 4 Mbyte, which includes memory space beyond the flash EEPROM memory.
7:9	—	Reserved
10:12	AT	Address type. This field can be used to specify that accesses involving the memory bank are limited to a certain address space type. These bits are used in conjunction with the ATM bits in the OR. The default value at reset is to map data only. For a full definition of address types, refer to <a href="#">9.5.7.6 Address Types</a> .
13:27	—	Reserved
28:30	DMCS	Dual-mapping chip select. This field determines which chip-select pin is assigned for dual mapping. 000 = CS[0] 001 = CS[1] 010 = CS[2] 011 = CS[3] 1xx = Reserved
31	DME	Dual mapping enabled. This bit indicates that the contents of the dual-mapping registers and associated base and option registers are valid and enables the dual-mapping operation. The default value at reset comes from the internal data bus that reflects the reset configuration word. See <a href="#">10.6 Dual Mapping of the Internal Flash EEPROM Array</a> for more information. 0 = Dual mapping is not active 1 = Dual mapping is active

### 10.8.6 Dual-Mapping Option Register

**DMOR — Dual-Mapping Option Register**

**0x2F C144**

MSB 0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
0	AM*						RESERVED			ATM			RESERVED		
RESET:															
0	0	0	0	0	0	0	0	0	0	0	0	1	0	0	0
16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	LSB 31
RESERVED															
RESET:															
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0

\*It is recommended that this field would hold values that are the power of 2 minus 1 (e.g.,  $2^3 - 1 = 7$  [0b111]).



**Table 10-10 DMOR Bit Descriptions**

Bit(s)	Name	Description
0	—	Reserved
1:6	AM	Address mask. The address mask field of each option register provides for masking any of the corresponding bits in the associated base register. By masking the address bits independently, external devices of different size address ranges can be used. Any clear bit masks the corresponding address bit. Any set causes the corresponding address bit to be used in the comparison with the address pins. Address mask bits can be set or cleared in any order in the field, allowing a resource to reside in more than one area of the address map. This field can be read or written at any time.
7:9	—	Reserved
10:12	ATM	<p>Address type mask. This field can be used to mask certain address type bits, allowing more than one address space type to be assigned to a chip select. Any set bit causes the corresponding address type code bits to be used as part of the address comparison. Any cleared bit masks the corresponding address type code bit.</p> <p>To instruct the memory controller to ignore address type codes as part of the address comparison, clear the ATM bits.</p> <p>Note: Following a system reset, the ATM bits are cleared in DMOR, except the ATM2 bit. This means that only data accesses are dual mapped. Refer to the address types definition in <a href="#">Table 9-8</a>.</p>
13:31	—	Reserved

